

Principles of Communications

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Chapter 10: Information Theory

Textbook: Chapter 12

Communication Systems Engineering: Ch 6.1, Ch
9.1~9.2

Information Theory

- **Information theory** is one of the key concepts in modern communications
- It deals with **fundamental limits** on communications
 - What is the highest rate at which information can be reliably transmitted over a communication channel?
 - What is the lowest rate at which information can be compressed and still be retrievable with small or no error?
 - What is the complexity of such optimal schemes?
- Topics to discuss
 - Modeling of information source
 - Source coding theorem
 - Modeling of communication channel
 - Channel capacity

10.1 Modeling of Information Source

- Information sources can be modeled by random processes
- The simplest model for information source is **discrete memoryless source (DMS)**, a **discrete-time, discrete-amplitude** random process with **i.i.d** random variables
- A full description of DMS is given by:
 - Alphabet set $\mathcal{A} = \{a_1, a_1, \dots, a_N\}$ where the random variable X takes its values
 - Probabilities $\{p_i\}_{i=1}^N$
- The information conveyed in different information sources can be different

Information

- How to give a quantitative measure of information?
- Examples:
 - “the sun will rise” \Rightarrow no information
 - “it will rain tomorrow” \Rightarrow some information
 - “Final exam will be canceled” \Rightarrow infinite information
- Information is connected with the elements of surprise, which is the **result of uncertainty**.
 - The **smaller the probability** of an event is, the **more information** the occurrence of that event will convey

Measure of Information

- The information I that a source event x can will convey and the probability of the event $P(x)$ satisfy:

1. $I = I[P(x)]$

2. $P(x) \downarrow \rightarrow I \uparrow$, vice versa

$$P(x) = 1, I = 0$$

3. Consider multiple independent events x_1, x_2, \dots

$$I[P(x_1)P(x_2) \dots] = I[P(x_1)] + I[P(x_2)] + \dots$$

- Definition (**self information of symbol x**):

$$I = \log_a \frac{1}{P(X)} = -\log_a P(X)$$

$$a = e \text{ nat} \quad a = 2 \text{ bit}$$

Entropy (熵)

- Consider a discrete source with N possible symbols
- Entropy $H(.)$ is defined as the **average amount of information** conveyed per symbol

$$H(X) \stackrel{\Delta}{=} E[I(x_j)] = \sum_{j=1}^N P(x_j) \log_2 \frac{1}{P(x_j)} \text{ (bit/symbol)}$$

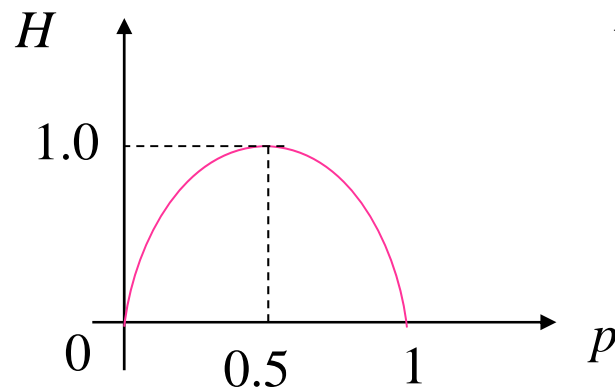
- Example: Consider a source having 3 symbols alphabet where $P(x_1) = 1/2$, $P(x_2) = P(x_3) = 1/4$, and symbols are statically independent. Determine the entropy of the source.

- Solution:
$$H = p_1 \log_2 \frac{1}{p_1} + p_2 \log_2 \frac{1}{p_2} + p_3 \log_2 \frac{1}{p_3}$$
$$= \frac{1}{2} \times 1 + \frac{1}{4} \times 2 + \frac{1}{4} \times 2 = 1.5 \text{ bit/Symbol}$$

Entropy (Cont'd)

- How to maximize entropy?
- Consider binary case with two symbol alphabet $\{0, 1\}$, if we let $P(1) = p$, and $P(0) = 1-p$, then

$$H = p \log_2 \frac{1}{p} + (1-p) \log_2 \frac{1}{1-p}$$



Entropy is maximized when all the symbols are equiprobable

- N symbols: $H = \sum_{n=1}^N \frac{1}{N} \log_2 N = \log_2 N$ bit/symbol

Exercise

- A source with bandwidth 4000Hz is sampled at the Nyquist rate. Assuming that the resulting sequence can be approximately modeled by a discrete memoryless source with alphabet $\{-2, -1, 0, 1, 2\}$ and with corresponding probabilities $\{1/2, 1/4, 1/8, 1/16, 1/16\}$, determine the rate of the source in bit/sec

Solution

- We have

$$\begin{aligned} H(X) &= \frac{1}{2} \log_2 2 + \frac{1}{4} \log_2 4 + \frac{1}{8} \log_2 8 + 2 \times \frac{1}{16} \log_2 16 \\ &= \frac{15}{8} \text{ bits/sample} \end{aligned}$$

- Since we have 8000 samples/sec the source produces information at a rate of 15kbits/sec.

Joint and Conditional Entropy

- When dealing with two or more random sources, exactly in the same way that joint and conditional probabilities are introduced, one can introduce joint and conditional entropies.

- The **joint entropy** of (X, Y) is defined as

$$H(X, Y) = -\sum_{x, y} p(x, y) \log p(x, y)$$

- The **conditional entropy** of X given Y is defined as

$$H(X | Y) = -\sum_{x, y} p(x, y) \log p(x | y)$$

- Using chain rule, it can be shown that

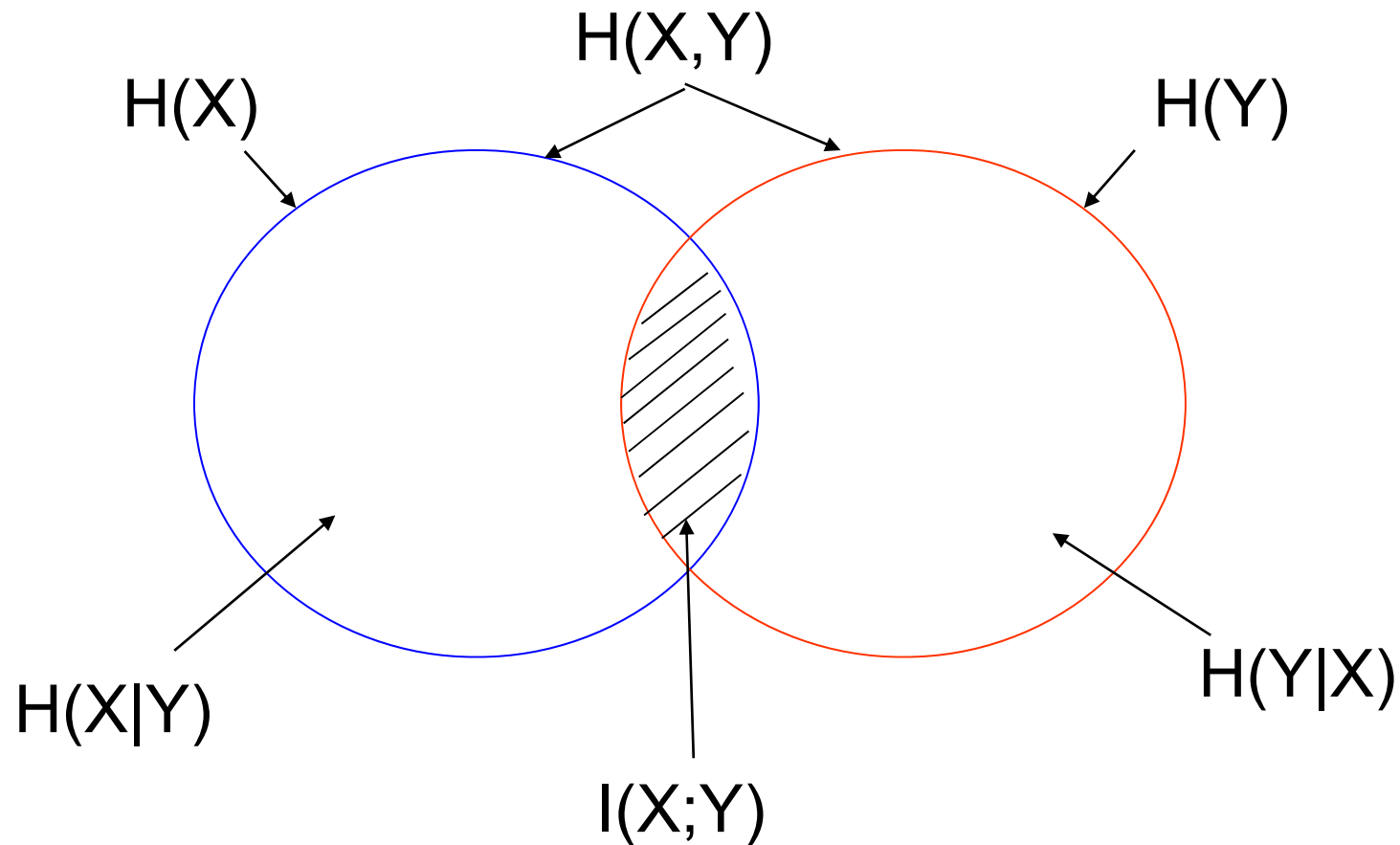
$$H(X, Y) = H(X | Y) + H(Y)$$

Mutual Information

- Given by
 - $H(X)$ denotes the uncertainty of the random variable X
 - $H(X|Y)$ denotes the uncertainty of random variable X after random variable Y is known.
- Then, $H(X)-H(X|Y)$
 - Denotes the amount of uncertainty of X that has been removed given Y is known
 - In other words, it is the amount of information provided by random variable Y about random variable X
- Definition of **mutual information**

$$I(X;Y) = H(X) - H(X | Y)$$

Entropy, Conditional Entropy and Mutual Information



Differential Entropy

- The **differential entropy** of a **discrete-time continuous alphabet source** X with pdf $f(x)$ is defined as:

$$h(X) = -\int_{-\infty}^{\infty} f_X(x) \log f_X(x) dx$$

- Example: the differential entropy of $X \sim N(0, \sigma^2)$ is

$$h(X) = \frac{1}{2} \log_2 (2\pi e \sigma^2) \text{ bits}$$

- **Mutual information** between two continuous random variables X and Y :

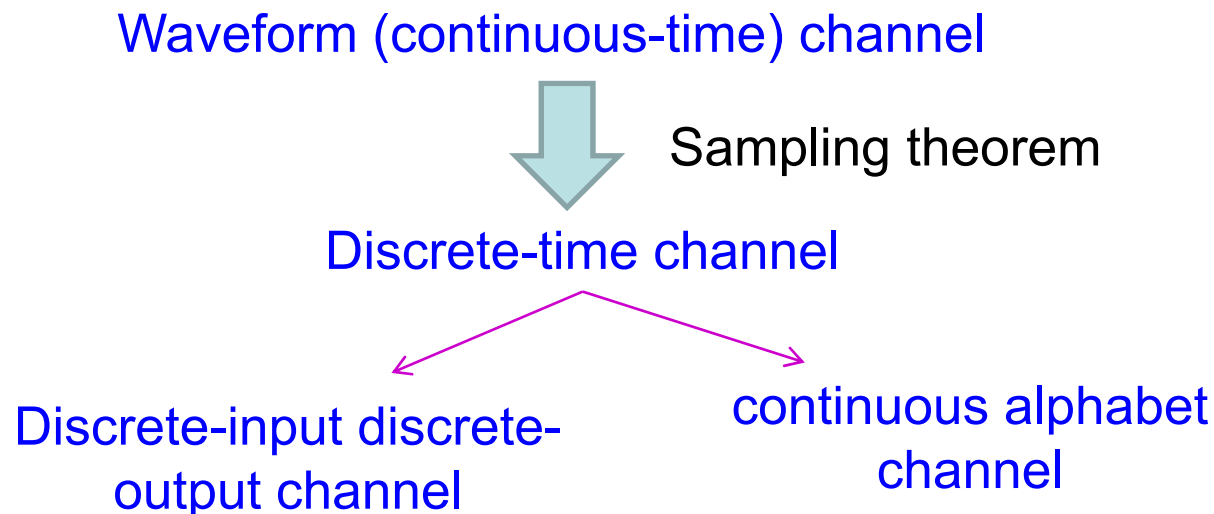
$$I(X; Y) = h(X) - h(X | Y)$$

10.2 Source Coding Theorem

- **Source coding theorem:**
 - A source with entropy (or entropy rate) H can be encoded with an arbitrarily small error probability at any rate R (bits/source output) as long as $R > H$.
 - Conversely, if $R < H$, the error probability will be bounded away from zero, independent of the complexity of the encoder and decoder employed

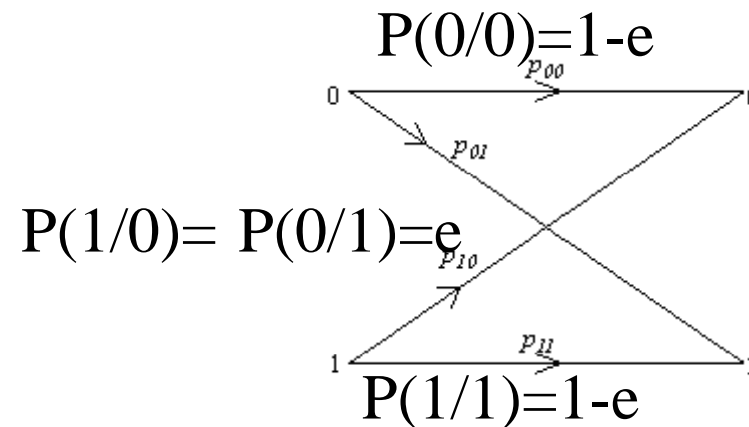
10.3 Modeling of Communication Channel

- Recall that a communication channel is any medium over which information can be transmitted
- It is characterized by a relationship between its input and output, which is generally a **stochastic relation** due to the presence of fading and noise



Binary-Symmetric Channel

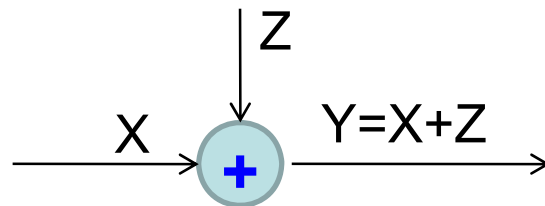
- BSC channel is characterized by the crossover probability $e = P(0|1) = P(1|0)$
- For instance, $e = Q\left(\sqrt{\frac{2E_b}{N_0}}\right)$



AWGN Channel

- Both input and output are real numbers
- The input satisfy some power constraint

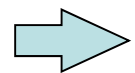
$$\sum_{i=1}^n x_i^2 \leq P$$



10.4 Channel Capacity

- In 1948, Shannon proved that
 - there exists a **maximum rate**, called **channel capacity** and denoted as C in bits/sec, at which one can communicate over a channel with arbitrarily small error probability
 - one can theoretically transmit over a channel at a rate $R \leq C$ with almost error free
 - Otherwise, if $R > C$, then **reliable transmission is not possible**
 - The capacity of a discrete-memoryless channel is given by

$$C = \max_{p(x)} I(X;Y) \text{ (max over all possible input distribution)}$$



The Noisy Channel Coding Theorem

(one of the fundamental results in information theory)

Claude E. Shannon (1916-2001)



Binary Symmetric Channel Capacity

- Since $I(X;Y) = H(Y) - H(Y | X)$
$$= H(Y) - \sum p(x)H(Y | X = x)$$
$$= H(Y) - \sum p(x)H(P_e)$$
$$= H(Y) - H(P_e)$$
$$\leq 1 - H(P_e)$$

- Here, $H(p) = -p \log_2 p - (1-p) \log_2 (1-p)$

$H(Y) \leq 1$ Equality holds when X is equal probably

- Thus, the capacity of a BSC is

$$C = 1 - H(P_e)$$

Gaussian Channel Capacity

- Consider a discrete-time Gaussian channel with

$$Y = X + Z$$

- Input power constraint: $\sum_{i=1}^n x_i^2 \leq P$
- $Z \sim N(0, P_N)$

- Its capacity is given by (proof?)

$$C = \frac{1}{2} \log \left(1 + \frac{P}{P_N} \right)$$


- Now, consider a continuous-time, bandlimited AWGN channel with noise PSD= $N_0/2$, input power constraint P , bandwidth W .
- Sample it at Nyquist rate and obtain a discrete-time channel. The power/sample will be P and the noise power/sample will be

$$P_N = \int_{-W}^W \frac{N_0}{2} df = WN_0$$

- Thus,

$$C = \frac{1}{2} \log \left(1 + \frac{P}{N_0 W} \right) \text{ bits/transmission}$$

- Since the number of transmissions/sec is $2W$, we obtain the channel capacity in bits/sec



$$C = W \log \left(1 + \frac{P}{N_0 W} \right) \text{ bits/sec}$$

(Shannon Formula)

Example

- Find the capacity of a telephone channel with bandwidth $W=3000\text{Hz}$, and SNR of 39dB
- Solution:
 - The SNR of 39 dB is equivalent to 7943. Using Shannon formula, we have

$$C = 3000 \log(1 + 7943) \approx \sim 38,867 \text{ bits/sec}$$

Insights from Shannon Formula

1. Increasing signal power P increases the capacity C
 - When SNR is high enough, every doubling of P adds additional B bits/s in capacity
 - When P approaches infinity, so is C
2. Increasing channel bandwidth W can increase C , but cannot increase infinitely (as noise power also increases)

$$\begin{aligned}\lim_{W \rightarrow \infty} C &= \lim_{W \rightarrow \infty} \left[\frac{WN_0}{P} \log \left(1 + \frac{P}{N_0W} \right) \right] \frac{P}{N_0} \\ &= \frac{P}{N_0} \log e = 1.44 \frac{P}{N_0}\end{aligned}$$

3. Bandwidth efficiency – energy efficiency tradeoff

- In any practical system, we must have

$$R \leq W \log_2 \left(1 + \frac{P}{N_0 W} \right)$$

- Defining $r=R/W$, the **spectral bit rate**

$$r = \frac{R}{W} \leq \log_2 \left(1 + \frac{P}{N_0 W} \right)$$

- Let E_b be the energy per bit, $E_b = \frac{P}{R}$

- Then, $r \leq \log_2 \left(1 + r \frac{E_b}{N_0} \right)$ $E_b/N_0 = \text{SNR per bit}$
 $r = \text{spectral efficiency}$

□ As $r=R/B \rightarrow 0$

$$\begin{aligned} \left. \frac{E_b}{N_0} \right|_{r \rightarrow 0} &= \lim_{r \rightarrow 0} \frac{1}{r} (2^r - 1) \\ &= \ln 2 \\ &= 0.693 \\ &= -1.59 \text{ dB} \end{aligned}$$

Shannon Limit, an absolute minimum for reliable communication

